

TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH RESET OUTPUT TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT FAST-TRANSIENT-RESPONSE 750-mA LOW-DROPOUT VOLTAGE REGULATORS

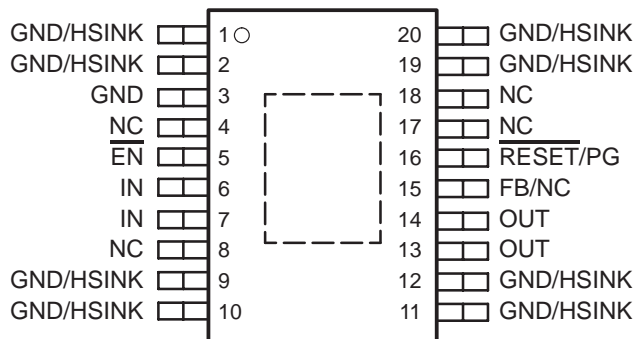
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- Open Drain Power-On Reset With 200-ms Delay (TPS777xx)
- Open Drain Power Good (TPS778xx)
- 750-mA Low-Dropout Voltage Regulator
- Available in 1.5-V, 1.8-V, 2.5-V, 3.3-V Fixed Output and Adjustable Versions
- Dropout Voltage to 260 mV (Typ) at 750 mA (TPS77x33)
- Ultralow 85 μA Typical Quiescent Current
- Fast Transient Response
- 2% Tolerance Over Specified Conditions for Fixed-Output Versions
- 8-Pin SOIC and 20-Pin TSSOP PowerPAD™ (PWP) Package
- Thermal Shutdown Protection

description

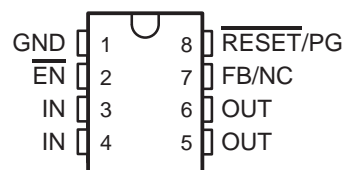
TPS777xx and TPS778xx are designed to have a fast transient response and be stable with a 10- μF low ESR capacitors. This combination provides high performance at a reasonable cost.

PWP PACKAGE
(TOP VIEW)

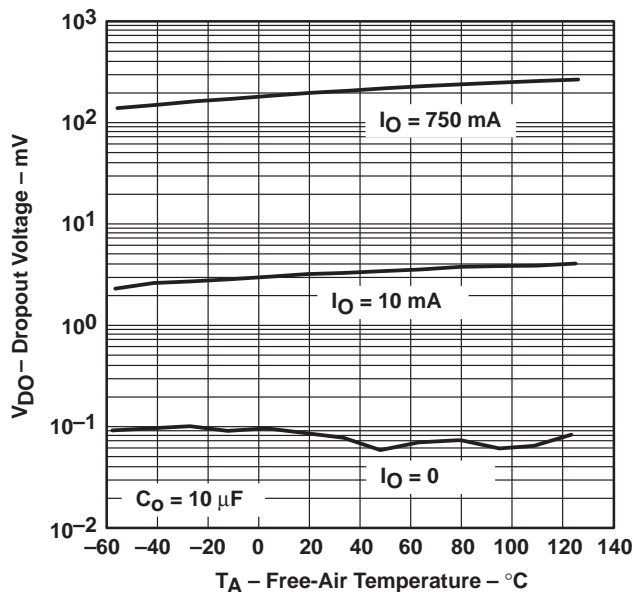


NC – No internal connection

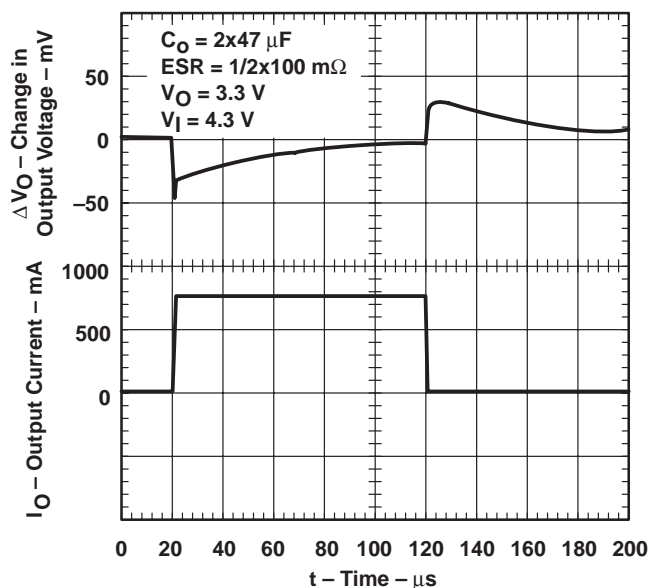
D PACKAGE
(TOP VIEW)



TPS77x33
DROPOUT VOLTAGE
vs
FREE-AIR TEMPERATURE



TPS77x33
LOAD TRANSIENT RESPONSE



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PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

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TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH $\overline{\text{RESET}}$ OUTPUT
TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
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description (continued)

Because the PMOS device behaves as a low-value resistor, the dropout voltage is very low (typically 260 mV at an output current of 750 mA for the TPS77x33) and is directly proportional to the output current. Additionally, since the PMOS pass element is a voltage-driven device, the quiescent current is very low and independent of output loading (typically 85 μA over the full range of output current, 0 mA to 750 mA). These two key specifications yield a significant improvement in operating life for battery-powered systems. This LDO family also features a sleep mode; applying a TTL high signal to $\overline{\text{EN}}$ (enable) shuts down the regulator, reducing the quiescent current to 1 μA at $T_J = 25^\circ\text{C}$.

The $\overline{\text{RESET}}$ output of the TPS777xx initiates a reset in microcomputer and microprocessor systems in the event of an undervoltage condition. An internal comparator in the TPS777xx monitors the output voltage of the regulator to detect an undervoltage condition on the regulated output voltage.

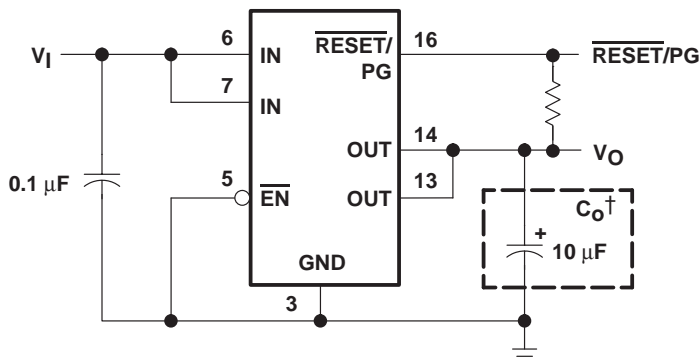
Power good (PG) of the TPS778xx is an active high output, which can be used to implement a power-on reset or a low-battery indicator.

The TPS777xx and TPS778xx are offered in 1.5-V, 1.8-V, 2.5-V, and 3.3-V fixed-voltage versions and in an adjustable version (programmable over the range of 1.5 V to 5.5 V for TPS77701 option and 1.2 V to 5.5 V for TPS77801 option). Output voltage tolerance is specified as a maximum of 2% over line, load, and temperature ranges. The TPS777xx and TPS778xx families are available in 8 pin SOIC and 20 pin PWP packages.

AVAILABLE OPTIONS

T_J	OUTPUT VOLTAGE (V)	PACKAGED DEVICES			
	TYP	TSSOP (PWP)		SOIC (D)	
-40°C to 125°C	3.3	TPS77733PWP	TPS77833PWP	TPS77733D	TPS77833D
	2.5	TPS77725PWP	TPS77825PWP	TPS77725D	TPS77825D
	1.8	TPS77718PWP	TPS77818PWP	TPS77718D	TPS77818D
	1.5	TPS77715PWP	TPS77815PWP	TPS77715D	TPS77815D
	Adjustable 1.5 V to 5.5 V	TPS77701PWP	—	TPS77701D	—
	Adjustable 1.2 V to 5.5 V	—	TPS77801PWP	—	TPS77801D

The TPS77x01 is programmable using an external resistor divider (see application information). The D and PWP packages are available taped and reeled. Add an R suffix to the device type (e.g., TPS77701DR).



† See application information section for capacitor selection details.

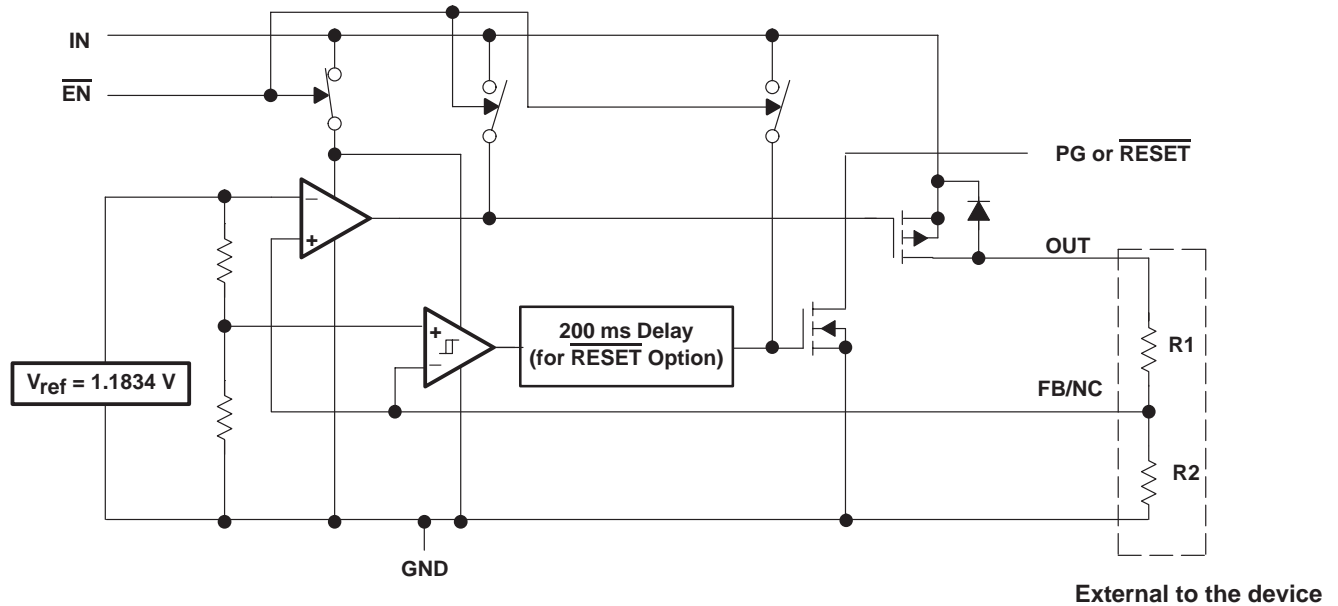
Figure 1. Typical Application Configuration for Fixed Output Options



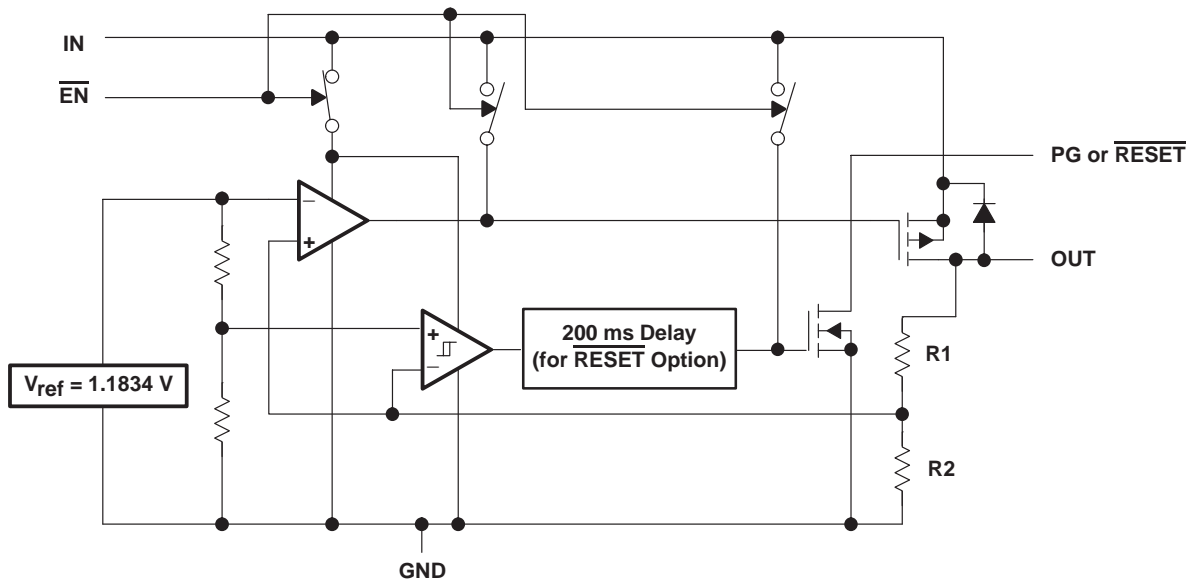
TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH $\overline{\text{RESET}}$ OUTPUT
 TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
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functional block diagram—adjustable version



functional block diagram—fixed-voltage version



TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH $\overline{\text{RESET}}$ OUTPUT
TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
FAST-TRANSIENT-RESPONSE 750-mA LOW-DROPOUT VOLTAGE REGULATORS

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Terminal Functions

SOIC Package (TPS777xx)

TERMINAL NAME	NO.	I/O	DESCRIPTION
$\overline{\text{EN}}$	2	I	Enable input
FB/NC	7	I	Feedback input voltage for adjustable device (no connect for fixed options)
GND	1		Regulator ground
IN	3, 4	I	Input voltage
OUT	5, 6	O	Regulated output voltage
$\overline{\text{RESET}}$	8	O	$\overline{\text{RESET}}$ output

TSSOP Package (TPS777xx)

TERMINAL NAME	NO.	I/O	DESCRIPTION
$\overline{\text{EN}}$	5	I	Enable input
FB/NC	15	I	Feedback input voltage for adjustable device (no connect for fixed options)
GND	3		Regulator ground
GND/HSINK	1, 2, 9, 10, 11, 12, 19, 20		Ground/heatsink
IN	6, 7	I	Input
NC	4, 8, 17, 18		No connect
OUT	13, 14	O	Regulated output voltage
$\overline{\text{RESET}}$	16	O	$\overline{\text{RESET}}$ output

SOIC Package (TPS778xx)

TERMINAL NAME	NO.	I/O	DESCRIPTION
$\overline{\text{EN}}$	2	I	Enable input
FB/NC	7	I	Feedback input voltage for adjustable device (no connect for fixed options)
GND	1		Regulator ground
IN	3, 4	I	Input voltage
OUT	5, 6	O	Regulated output voltage
PG	8	O	PG output

TSSOP Package (TPS778xx)

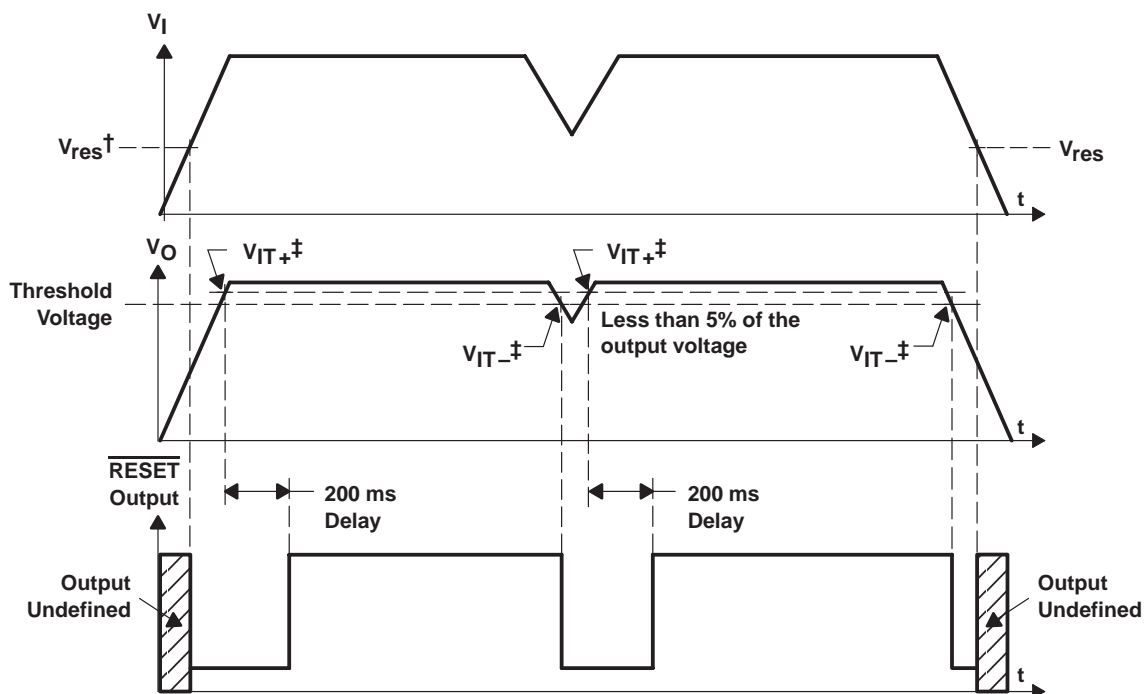
TERMINAL NAME	NO.	I/O	DESCRIPTION
$\overline{\text{EN}}$	5	I	Enable input
FB/NC	15	I	Feedback input voltage for adjustable device (no connect for fixed options)
GND	3		Regulator ground
GND/HSINK	1, 2, 9, 10, 11, 12, 19, 20		Ground/heatsink
IN	6, 7	I	Input
NC	4, 8, 17, 18		No connect
OUT	13, 14	O	Regulated output voltage
PG	16	O	PG output



TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH $\overline{\text{RESET}}$ OUTPUT
 TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
 FAST-TRANSIENT-RESPONSE 750-mA LOW-DROPOUT VOLTAGE REGULATORS

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TPS777xx $\overline{\text{RESET}}$ timing diagram



$^\dagger V_{res}$ is the minimum input voltage for a valid $\overline{\text{RESET}}$. The symbol V_{res} is not currently listed within EIA or JEDEC standards for semiconductor symbology.

$^\ddagger V_{IT-}$ – Trip voltage is typically 5% lower than the output voltage ($95\%V_o$). V_{IT-} to V_{IT+} is the hysteresis voltage.

TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH RESET OUTPUT
TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Input voltage range‡, V_I	–0.3 V to 13.5 V
Voltage range at \overline{EN}	–0.3 V to 16.5 V
Maximum \overline{RESET} voltage (TPS777xx)	16.5 V
Maximum PG voltage (TPS778xx)	16.5 V
Peak output current	Internally limited
Output voltage, V_O (OUT, FB)	7 V
Continuous total power dissipation	See dissipation rating tables
Operating virtual junction temperature range, T_J	–40°C to 125°C
Storage temperature range, T_{stg}	–65°C to 150°C
ESD rating, HBM	2 kV

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

‡ All voltage values are with respect to network terminal ground.

DISSIPATION RATING TABLE 1 – FREE-AIR TEMPERATURES

PACKAGE	AIR FLOW (CFM)	$T_A < 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
D	0	568 mW	5.68 mW/°C	312 mW	227 mW
	250	904 mW	9.04 mW/°C	497 mW	361 mW

DISSIPATION RATING TABLE 2 – FREE-AIR TEMPERATURES

PACKAGE	AIR FLOW (CFM)	$T_A < 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
PWP§	0	2.9 W	23.5 mW/°C	1.9 W	1.5 W
	300	4.3 W	34.6 mW/°C	2.8 W	2.2 W
PWP¶	0	3 W	23.8 mW/°C	1.9 W	1.5 W
	300	7.2 W	57.9 mW/°C	4.6 W	3.8 W

§ This parameter is measured with the recommended copper heat sink pattern on a 1-layer PCB, 5-in × 5-in PCB, 1 oz. copper, 2-in × 2-in coverage (4 in²).

¶ This parameter is measured with the recommended copper heat sink pattern on a 8-layer PCB, 1.5-in × 2-in PCB, 1 oz. copper with layers 1, 2, 4, 5, 7, and 8 at 5% coverage (0.9 in²) and layers 3 and 6 at 100% coverage (6 in²). For more information, refer to TI technical brief SLMA002.

recommended operating conditions

		MIN	MAX	UNIT
Input voltage, V_I #		2.7	10	V
Output voltage range, V_O	TPS77701	1.5	5.5	V
	TPS77801	1.2	5.5	
Operating virtual junction temperature, T_J (see Note 1)		–40	125	°C

To calculate the minimum input voltage for your maximum output current, use the following equation: $V_{I(\min)} = V_{O(\max)} + V_{DO(\max \text{ load})}$.



TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH RESET OUTPUT
TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
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electrical characteristics over recommended operating temperature range (T_J = –40°C to 125°C),
V_I = V_O(typ) + 1 V, I_O = 1 mA, EN = 0 V, C_O = 10 μF (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT		
Output voltage (10 μA to 750 mA load) (see Note 2)	TPS77701	1.5 V ≤ V _O ≤ 5.5 V, T _J = 25°C	V _O			V		
		1.5 V ≤ V _O ≤ 5.5 V,	0.98V _O	1.02V _O				
	TPS77801	1.2 V ≤ V _O ≤ 5.5 V, T _J = 25°C	V _O					
		1.2 V ≤ V _O ≤ 5.5 V,	0.98V _O	1.02V _O				
	TPS77x15	T _J = 25°C, 2.7 V < V _{IN} < 10 V	1.5					
		T _J = –40°C to 125°C, 2.7 V < V _{IN} < 10 V	1.470	1.530				
	TPS77x18	T _J = 25°C, 2.8 V < V _{IN} < 10 V	1.8					
		T _J = –40°C to 125°C, 2.8 V < V _{IN} < 10 V	1.764	1.836				
	TPS77x25	T _J = 25°C, 3.5 V < V _{IN} < 10 V	2.5					
		T _J = –40°C to 125°C, 3.5 V < V _{IN} < 10 V	2.450	2.550				
	TPS77x33	T _J = 25°C, 4.3 V < V _{IN} < 10 V	3.3					
		T _J = –40°C to 125°C, 4.3 V < V _{IN} < 10 V	3.234	3.366				
Quiescent current (GND current) (see Note 2)		10 μA < I _O < 750 mA, T _J = 25°C	85			μA		
		I _O = 750 mA,	125					
Output voltage line regulation (ΔV _O /V _O) (see Notes 2 and 3)		V _O + 1 V < V _I ≤ 10 V, T _J = 25°C	0.01			%/V		
Load regulation			3			mV		
Output noise voltage (TPS77x18)		BW = 200 Hz to 100 kHz, C _O = 10 μF, T _J = 25°C, I _C = 750 μA	54			μVrms		
Output current limit		V _O = 0 V	1.7	2		A		
Thermal shutdown junction temperature			150			°C		
Standby current		EN = V _I , T _J = 25°C, 2.7 V < V _I < 10 V	1			μA		
		EN = V _I , 2.7 V < V _I < 10 V	10			μA		
FB input current	TPS77x01	FB = 1.5 V	2			nA		
High level enable input voltage			1.7			V		
Low level enable input voltage			0.9			V		
Power supply ripple rejection (see Note 2)		f = 1 KHz, C _O = 10 μF, T _J = 25°C	60			dB		
Reset (TPS77xx)	Minimum input voltage for valid RESET		I _O (RESET) = 300 μA			V		
	Trip threshold voltage		V _O decreasing			92	98	%V _O
	Hysteresis voltage		Measured at V _O			0.5		%V _O
	Output low voltage		V _I = 2.7 V, I _O (RESET) = 1 mA	0.15	0.4		V	
	Leakage current		V(RESET) = 5 V	1			μA	
	RESET time-out delay			200			ms	

NOTES: 1. Minimum IN operating voltage is 2.7 V or V_O(typ) + 1 V, whichever is greater. Maximum IN voltage 10V.
 2. If V_O ≤ 1.8 V then V_{Imin} = 2.7 V, V_{Imax} = 10 V:

$$\text{Line Reg. (mV)} = (\%/V) \times \frac{V_O(V_{I\max} - 2.7 V)}{100} \times 1000$$

If V_O ≥ 2.5 V then V_{Imin} = V_O + 1 V, V_{Imax} = 10 V:

$$\text{Line Reg. (mV)} = (\%/V) \times \frac{V_O(V_{I\max} - (V_O + 1 V))}{100} \times 1000$$



TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH RESET OUTPUT
TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
FAST-TRANSIENT-RESPONSE 750-mA LOW-DROPOUT VOLTAGE REGULATORS

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electrical characteristics over recommended operating temperature range ($T_J = -40^{\circ}\text{C}$ to 125°C), $V_I = V_{O(\text{typ})} + 1\text{ V}$, $I_O = 1\text{ mA}$, $\overline{\text{EN}} = 0\text{ V}$, $C_O = 10\text{ }\mu\text{F}$ (unless otherwise noted) (continued)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
PG (TPS778xx)	Minimum input voltage for valid PG	$I_{O(\text{PG})} = 300\text{ }\mu\text{A}$		1.1		V
	Trip threshold voltage	V_O decreasing	92		98	% V_O
	Hysteresis voltage	Measured at V_O		0.5		% V_O
	Output low voltage	$V_I = 2.7\text{ V}$, $I_{O(\text{PG})} = 1\text{ mA}$		0.15	0.4	V
	Leakage current	$V_{(\text{PG})} = 5\text{ V}$			1	μA
Input current ($\overline{\text{EN}}$)		$\overline{\text{EN}} = 0\text{ V}$	-1	0	1	μA
		$\overline{\text{EN}} = V_I$	-1		1	
Dropout voltage (see Note 4)	TPS77733	$I_O = 750\text{ mA}$, $T_J = 25^{\circ}\text{C}$		260		mV
		$I_O = 750\text{ mA}$,			427	
	TPS77833	$I_O = 750\text{ mA}$, $T_J = 25^{\circ}\text{C}$		260		
		$I_O = 750\text{ mA}$,			427	

NOTE 3: I_N voltage equals $V_{O(\text{typ})} - 100\text{ mV}$; TPS77x01 output voltage set to 3.3 V nominal with external resistor divider. TPS77x15, TPS77x18, and TPS77x25 dropout voltage limited by input voltage range limitations (i.e., TPS77x33 input voltage needs to drop to 3.2 V for purpose of this test).

TYPICAL CHARACTERISTICS

Table of Graphs

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	Power supply ripple rejection	vs Frequency	9
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Z_O	Output impedance	vs Frequency	11
V_{DO}	Dropout voltage	vs Input voltage	12
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V_O	Output voltage	vs Time	19
	Equivalent series resistance (ESR)	vs Output current	21 – 24



TYPICAL CHARACTERISTICS

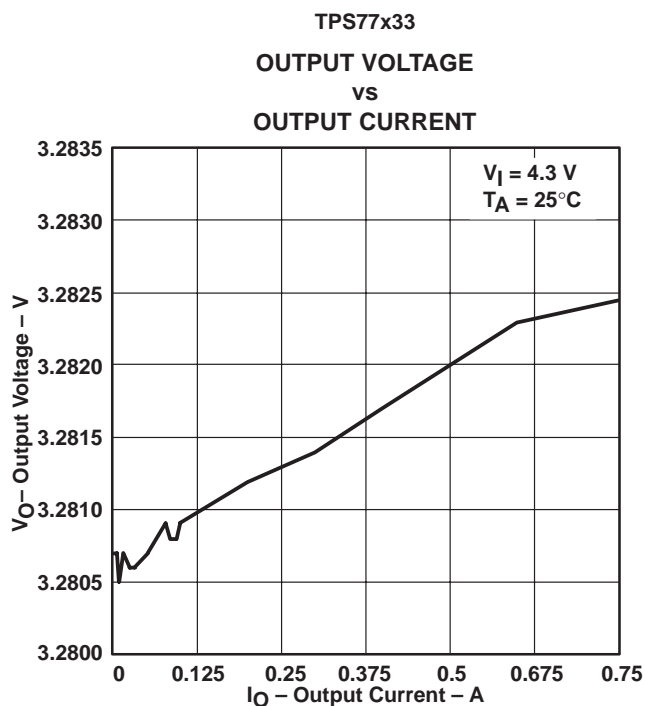


Figure 2

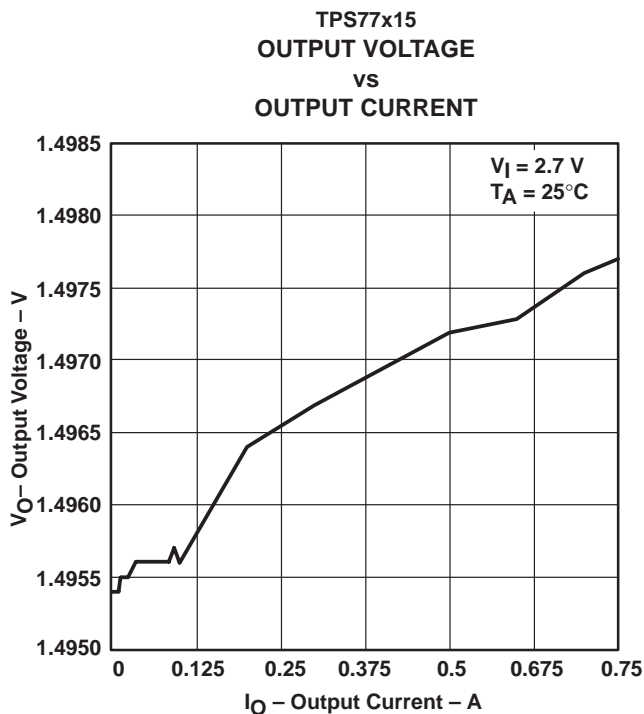


Figure 3

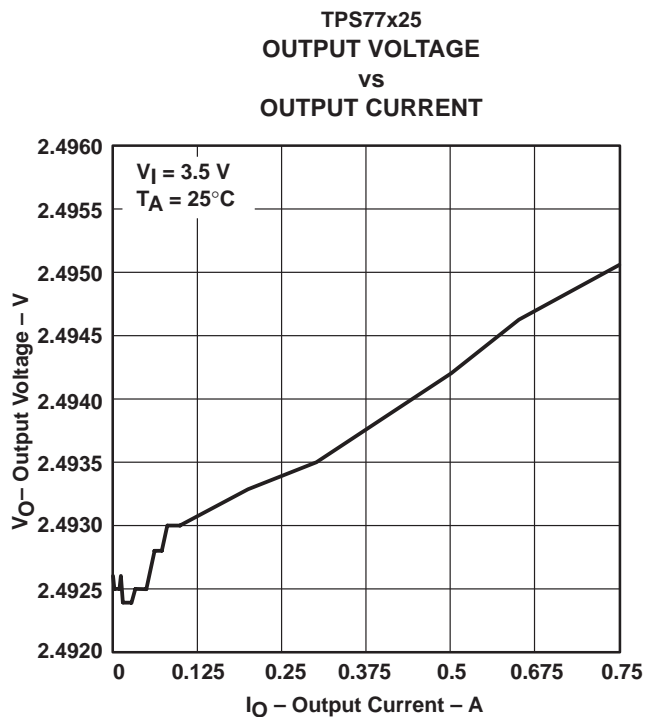


Figure 4

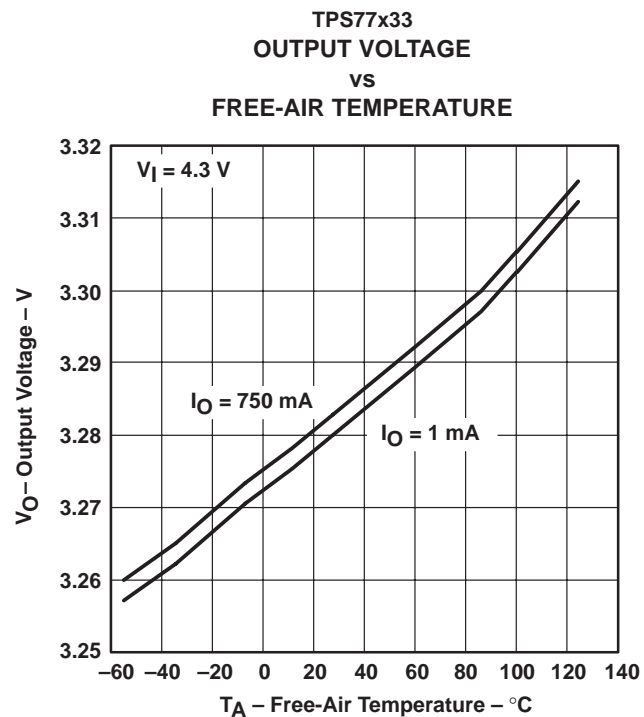


Figure 5

TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH $\overline{\text{RESET}}$ OUTPUT
 TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
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TYPICAL CHARACTERISTICS

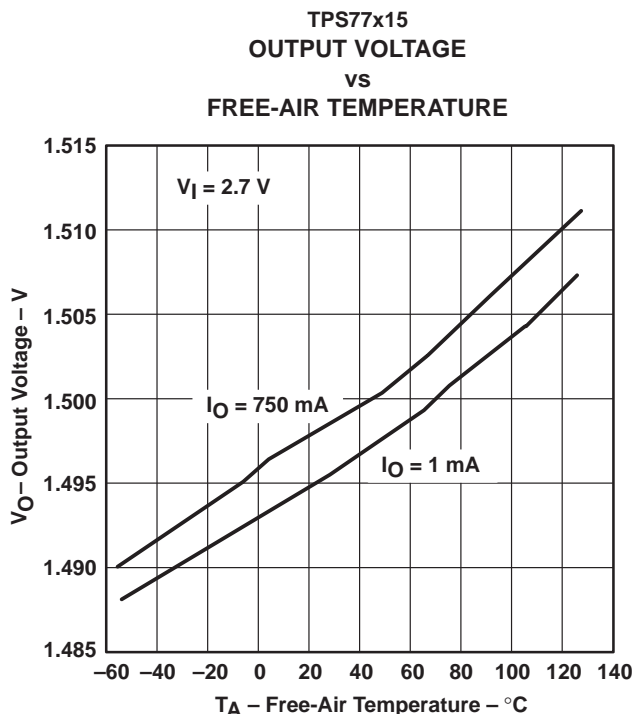


Figure 6

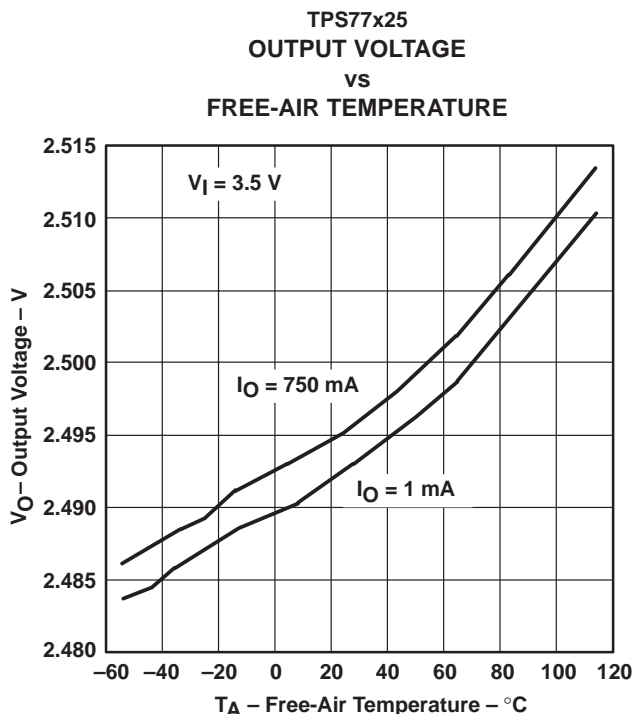


Figure 7

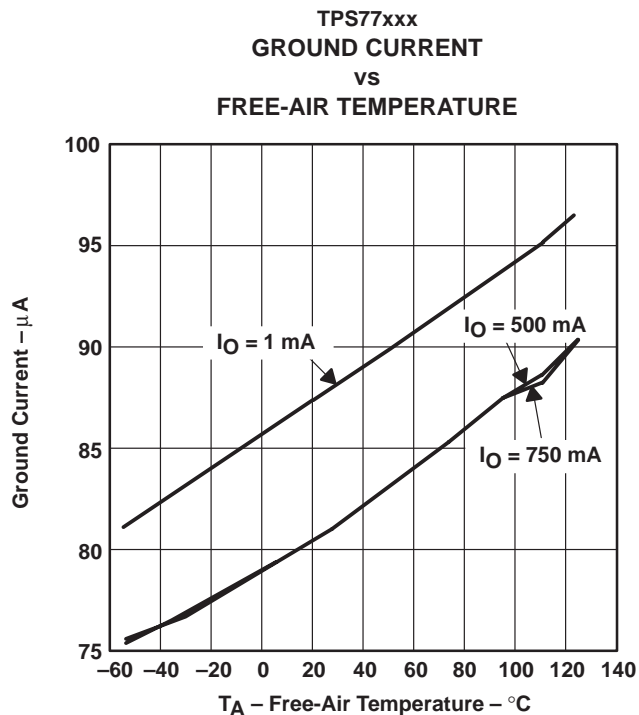


Figure 8

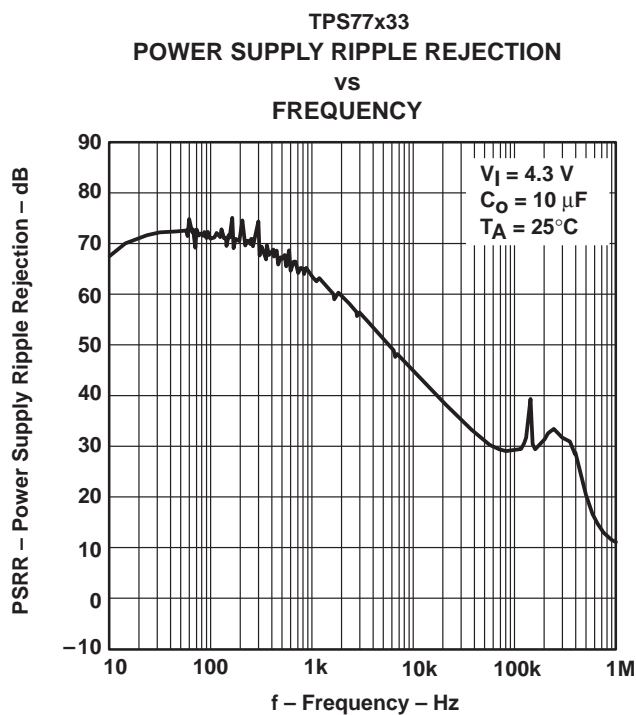


Figure 9



TYPICAL CHARACTERISTICS

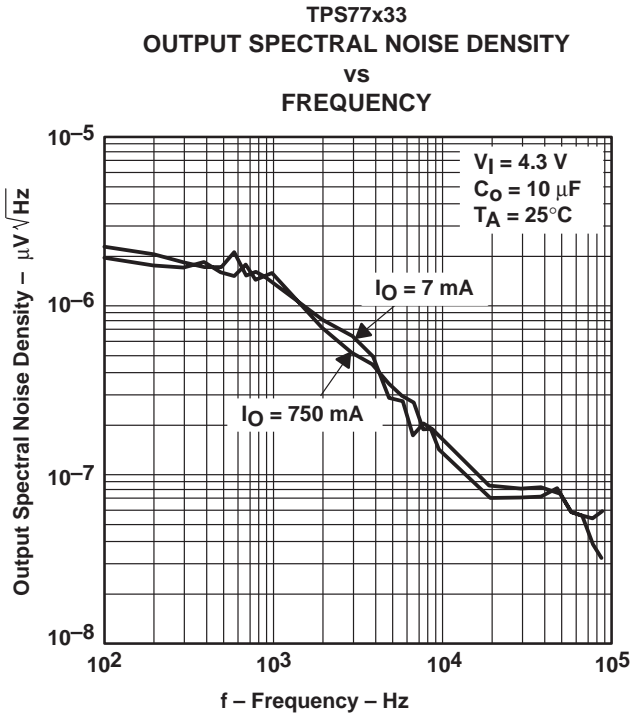


Figure 10

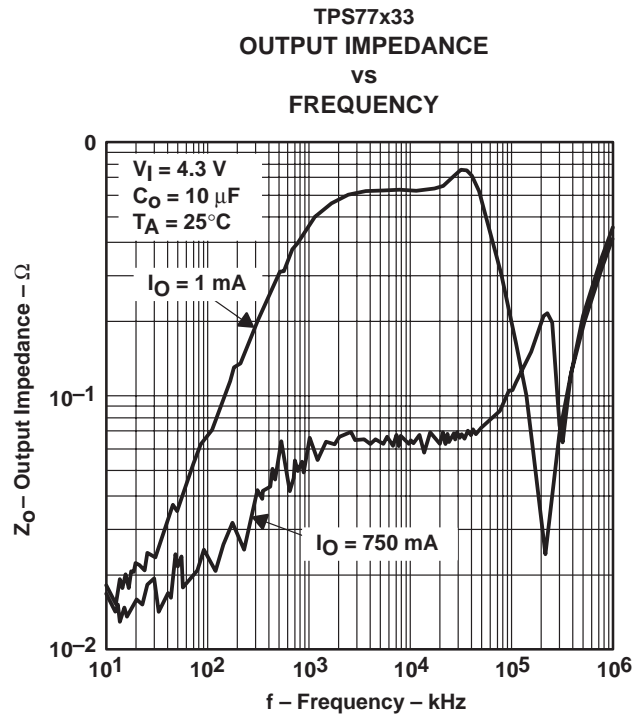


Figure 11

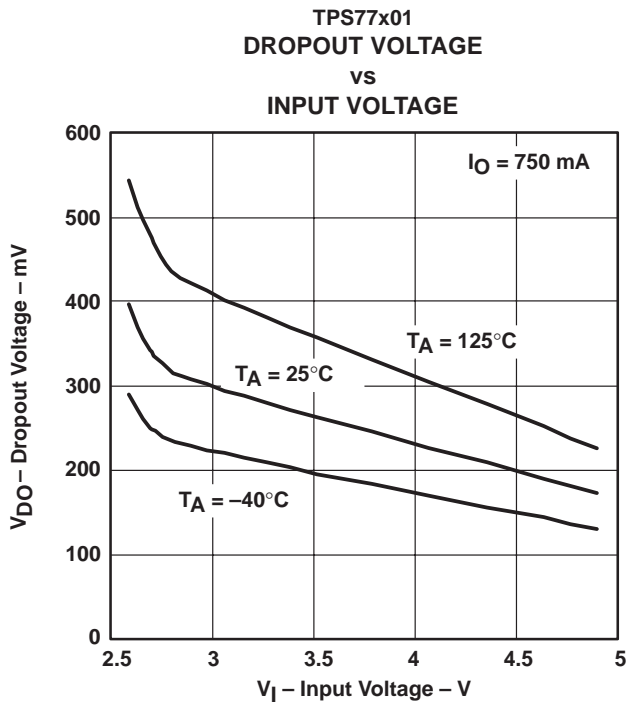


Figure 12

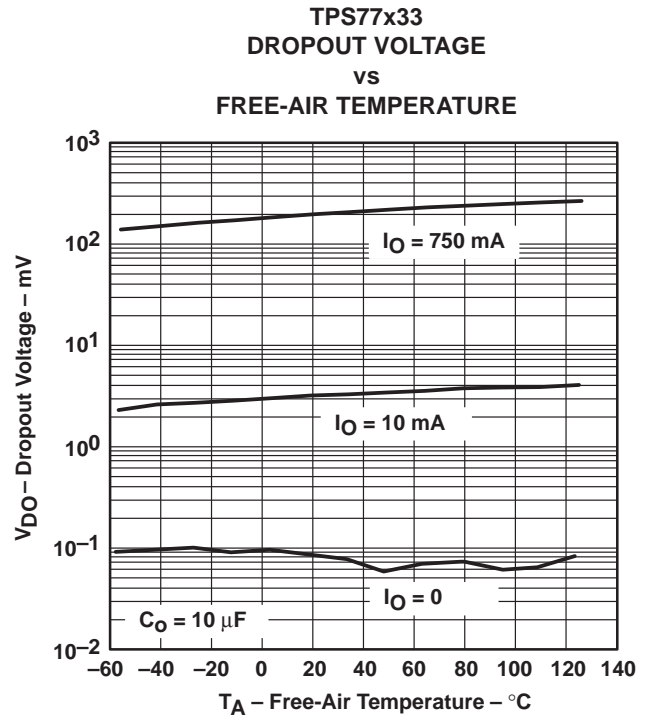


Figure 13

TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH $\overline{\text{RESET}}$ OUTPUT
 TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
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TYPICAL CHARACTERISTICS

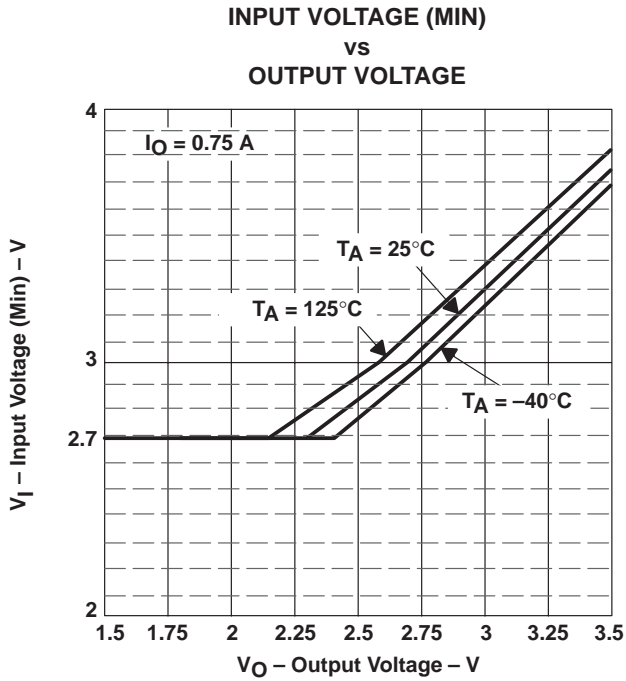


Figure 14

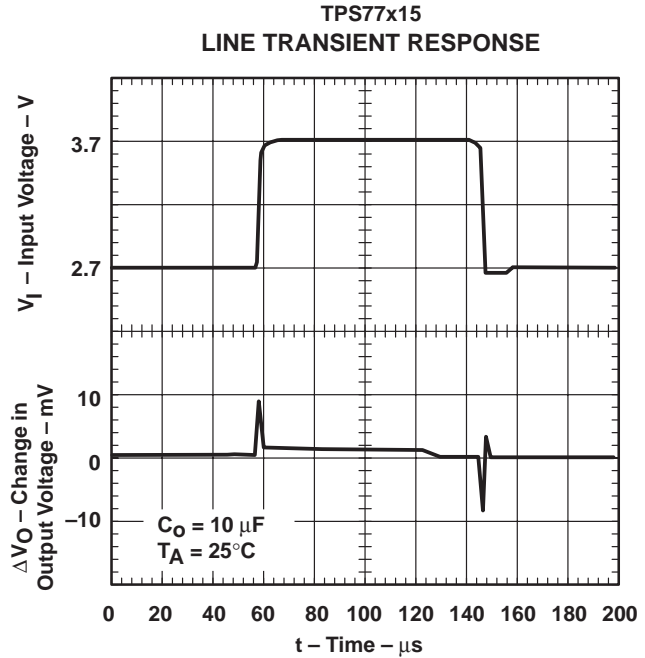


Figure 15

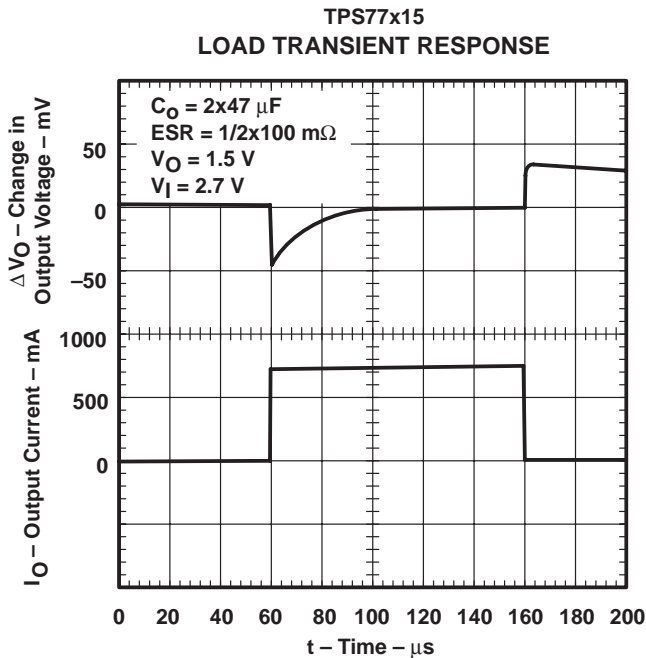


Figure 16

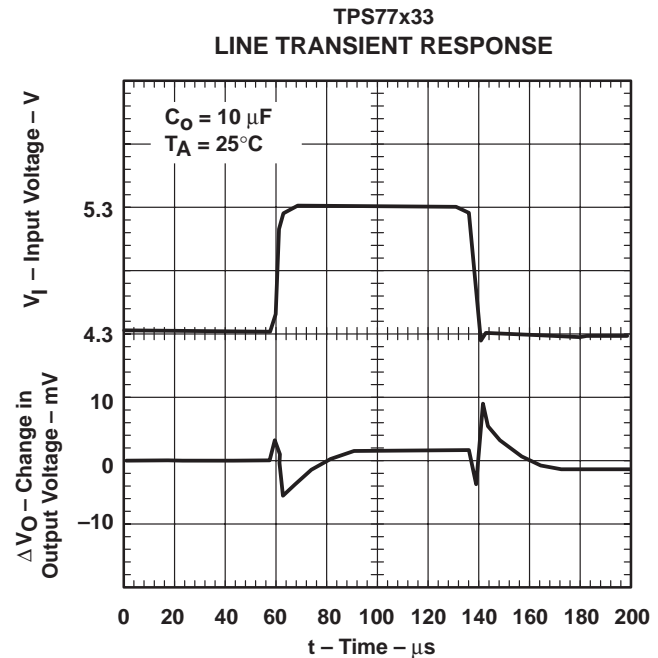


Figure 17



TYPICAL CHARACTERISTICS

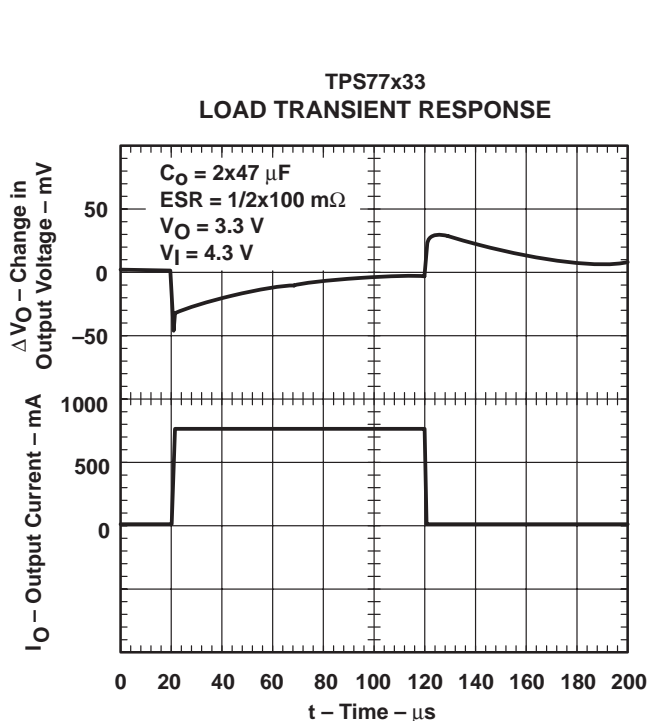


Figure 18

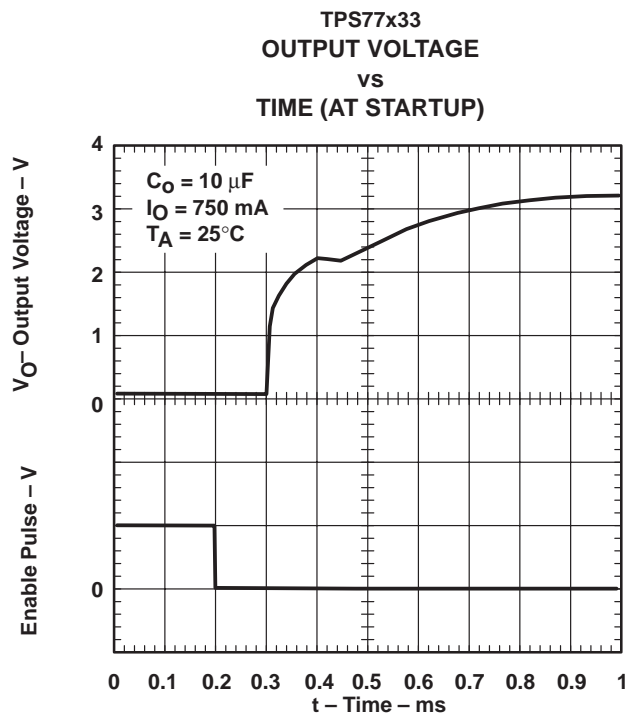


Figure 19

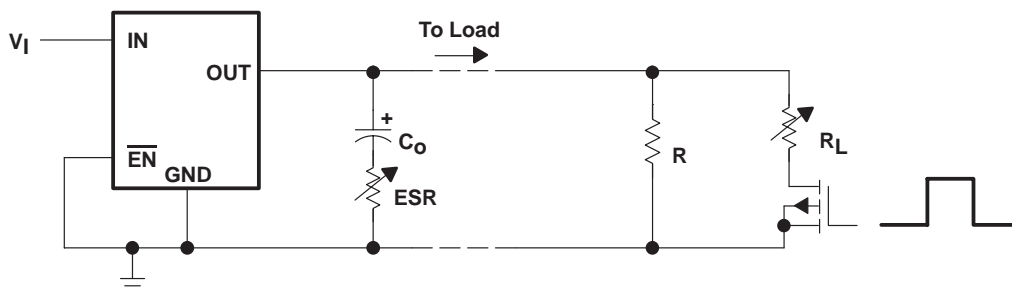


Figure 20. Test Circuit for Typical Regions of Stability (Figures 21 through 24) (Fixed Output Options)

TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH $\overline{\text{RESET}}$ OUTPUT
 TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT
 FAST-TRANSIENT-RESPONSE 750-mA LOW-DROPOUT VOLTAGE REGULATORS

SLVS230F – SEPTEMBER 1999 – REVISED JUNE 2003

TYPICAL CHARACTERISTICS

TYPICAL REGION OF STABILITY
 EQUIVALENT SERIES RESISTANCE†
 vs
 OUTPUT CURRENT

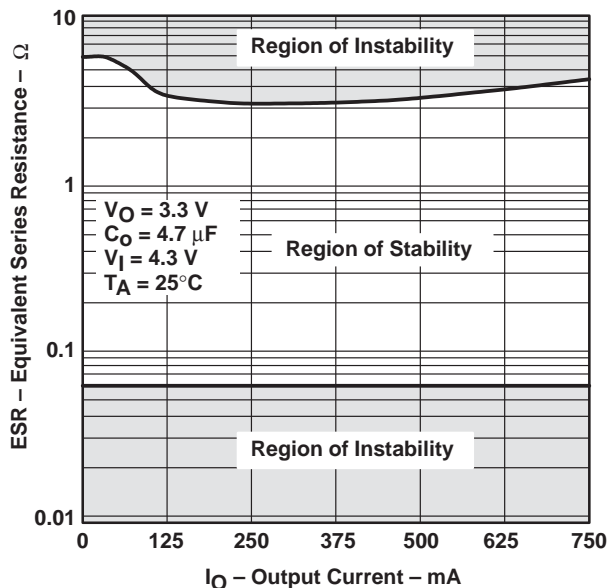


Figure 21

TYPICAL REGION OF STABILITY
 EQUIVALENT SERIES RESISTANCE†
 vs
 OUTPUT CURRENT

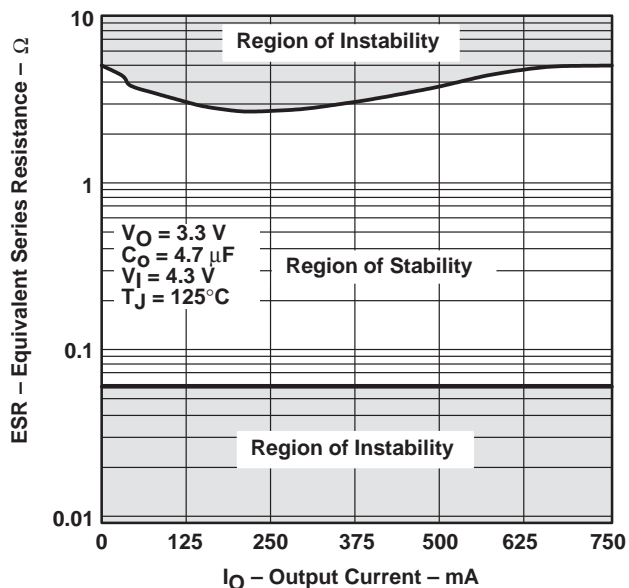


Figure 22

TYPICAL REGION OF STABILITY
 EQUIVALENT SERIES RESISTANCE†
 vs
 OUTPUT CURRENT

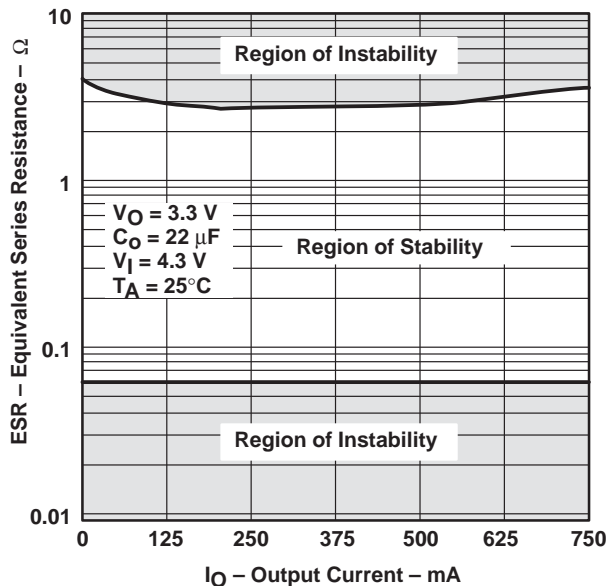


Figure 23

TYPICAL REGION OF STABILITY
 EQUIVALENT SERIES RESISTANCE†
 vs
 OUTPUT CURRENT

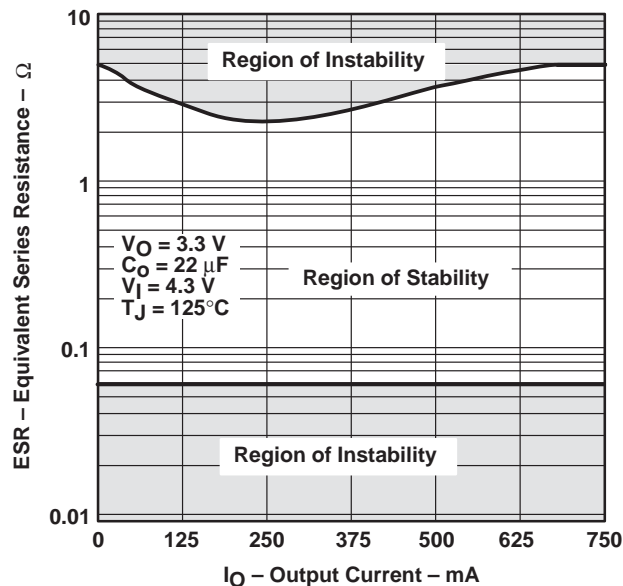


Figure 24

† Equivalent series resistance (ESR) refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to C_O .

TPS77701, TPS77715, TPS77718, TPS77725, TPS77733 WITH RESET OUTPUT TPS77801, TPS77815, TPS77818, TPS77825, TPS77833 WITH PG OUTPUT FAST-TRANSIENT-RESPONSE 750-mA LOW-DROPOUT VOLTAGE REGULATORS

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APPLICATION INFORMATION

The TPS777xx and TPS778xx families include four fixed-output voltage regulators (1.5 V, 1.8 V, 2.5 V, and 3.3 V), and an adjustable regulator, the TPS77x01 (adjustable from 1.5 V to 5.5 V for TPS77701 option and 1.2 V to 5.5 V for TPS77801 option).

device operation

The TPS777xx and TPS778xx feature very low quiescent current, which remains virtually constant even with varying loads. Conventional LDO regulators use a pnp pass element, the base current of which is directly proportional to the load current through the regulator ($I_B = I_C/\beta$). The TPS777xx and TPS778xx use a PMOS transistor to pass current; because the gate of the PMOS is voltage driven, operating current is low and invariable over the full load range.

Another pitfall associated with the pnp-pass element is its tendency to saturate when the device goes into dropout. The resulting drop in β forces an increase in I_B to maintain the load. During power up, this translates to large start-up currents. Systems with limited supply current may fail to start up. In battery-powered systems, it means rapid battery discharge when the voltage decays below the minimum required for regulation. The TPS777xx and TPS778xx quiescent currents remain low even when the regulator drops out, eliminating both problems.

The TPS777xx and TPS778xx families also feature a shutdown mode that places the output in the high-impedance state (essentially equal to the feedback-divider resistance) and reduces quiescent current to 2 μ A. If the shutdown feature is not used, \overline{EN} should be tied to ground.

minimum load requirements

The TPS777xx and TPS778xx families are stable even at zero load; no minimum load is required for operation.

FB—pin connection (adjustable version only)

The FB pin is an input pin to sense the output voltage and close the loop for the adjustable option. The output voltage is sensed through a resistor divider network to close the loop as it is shown in Figure 26. Normally, this connection should be as short as possible; however, the connection can be made near a critical circuit to improve performance at that point. Internally, FB connects to a high-impedance wide-bandwidth amplifier and noise pickup feeds through to the regulator output. Routing the FB connection to minimize/avoid noise pickup is essential.

external capacitor requirements

An input capacitor is not usually required; however, a ceramic bypass capacitor (0.047 μ F or larger) improves load transient response and noise rejection if the TPS777xx or TPS778xx are located more than a few inches from the power supply. A higher-capacitance electrolytic capacitor may be necessary if large (hundreds of milliamps) load transients with fast rise times are anticipated.

Like all low dropout regulators, the TPS777xx and TPS778xx require an output capacitor connected between OUT and GND to stabilize the internal control loop. The minimum recommended capacitance value is 10 μ F and the ESR (equivalent series resistance) must be between 50 m Ω and 1.5 Ω . Capacitor values 10 μ F or larger are acceptable, provided the ESR is less than 1.5 Ω . Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all suitable, provided they meet the requirements described previously.

APPLICATION INFORMATION

external capacitor requirements (continued)

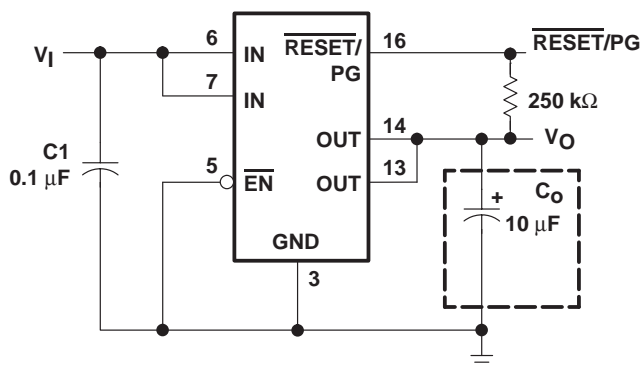


Figure 25. Typical Application Circuit (Fixed Versions)

programming the TPS77x01 adjustable LDO regulator

The output voltage of the TPS77x01 adjustable regulator is programmed using an external resistor divider as shown in Figure 26. The output voltage is calculated using:

$$V_O = V_{ref} \times \left(1 + \frac{R1}{R2}\right) \tag{1}$$

Where:

$$V_{ref} = 1.1834 \text{ V typ (the internal reference voltage)}$$

Resistors R1 and R2 should be chosen for approximately 10-μA divider current. Lower value resistors can be used but offer no inherent advantage and waste more power. Higher values should be avoided as leakage currents at FB increase the output voltage error. The recommended design procedure is to choose R2 = 110 kΩ to set the divider current at approximately 10 μA and then calculate R1 using:

$$R1 = \left(\frac{V_O}{V_{ref}} - 1\right) \times R2 \tag{2}$$

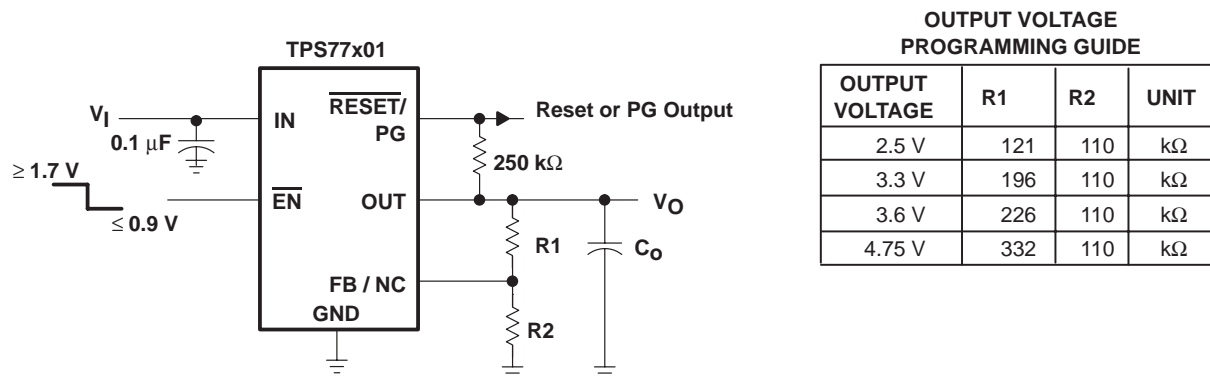


Figure 26. TPS77x01 Adjustable LDO Regulator Programming

APPLICATION INFORMATION

reset indicator

The TPS777xx features a $\overline{\text{RESET}}$ output that can be used to monitor the status of the regulator. The internal comparator monitors the output voltage: when the output drops to between 92% and 98% of its nominal regulated value, the $\overline{\text{RESET}}$ output transistor turns on, taking the signal low. The open-drain output requires a pullup resistor. If not used, it can be left floating. $\overline{\text{RESET}}$ can be used to drive power-on reset circuitry or as a low-battery indicator. $\overline{\text{RESET}}$ does not assert itself when the regulated output voltage falls outside the specified 2% tolerance, but instead reports an output voltage low relative to its nominal regulated value (refer to timing diagram for start-up sequence).

power-good indicator

The TPS778xx features a power-good (PG) output that can be used to monitor the status of the regulator. The internal comparator monitors the output voltage: when the output drops to between 92% and 98% of its nominal regulated value, the PG output transistor turns on, taking the signal low. The open-drain output requires a pullup resistor. If not used, it can be left floating. PG can be used to drive power-on reset circuitry or used as a low-battery indicator.

regulator protection

The TPS777xx and TPS778xx PMOS-pass transistors have a built-in back diode that conducts reverse currents when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. When extended reverse voltage is anticipated, external limiting may be appropriate.

The TPS777xx and TPS778xx also feature internal current limiting and thermal protection. During normal operation, the TPS777xx and TPS778xx limit output current to approximately 1.7 A. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds 150°C(typ), thermal-protection circuitry shuts it down. Once the device has cooled below 130°C(typ), regulator operation resumes.

APPLICATION INFORMATION

power dissipation and junction temperature

Specified regulator operation is assured to a junction temperature of 125°C; the maximum junction temperature should be restricted to 125°C under normal operating conditions. This restriction limits the power dissipation the regulator can handle in any given application. To ensure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation, $P_{D(max)}$, and the actual dissipation, P_D , which must be less than or equal to $P_{D(max)}$.

The maximum-power-dissipation limit is determined using the following equation:

$$P_{D(max)} = \frac{T_{Jmax} - T_A}{R_{\theta JA}}$$

Where:

T_{Jmax} is the maximum allowable junction temperature.

$R_{\theta JA}$ is the thermal resistance junction-to-ambient for the package, and is calculated as

$$\frac{1}{\text{derating factor}} \quad \text{from the dissipation rating tables.}$$

T_A is the ambient temperature.

The regulator dissipation is calculated using:

$$P_D = (V_I - V_O) \times I_O$$

Power dissipation resulting from quiescent current is negligible. Excessive power dissipation will trigger the thermal protection circuit.

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